General Description

The MAX1818 low-dropout linear regulator operates from a +2.5V to +5.5V supply and delivers a guaranteed 500mA load current with low 120mV dropout. The high-accuracy (\pm 1%) output voltage is preset at an internally trimmed voltage (see *Selector Guide*) or can be adjusted from 1.25V to 5.0V with an external resistive divider.

An internal PMOS pass transistor allows the low 125µA supply current to remain independent of load, making this device ideal for portable battery-operated equipment such as personal digital assistants (PDAs), cellular phones, cordless phones, base stations, and notebook computers.

Other features include an active-low open-drain reset output that indicates when the output is out of regulation, a 0.1µA shutdown, short-circuit protection, and thermal shutdown protection. The device is available in a miniature 800mW 6-pin SOT23 package.

Applications

Notebook Computers

Cellular and Cordless Telephones

Personal Digital Assistants (PDAs)

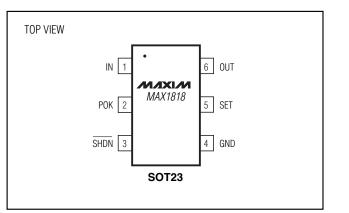
Palmtop Computers

Base Stations

USB Hubs

Docking Stations

Pin Configuration



Features

- ±1% Output Voltage Accuracy Preset at 1.5V, 1.8V, 2.0V, 2.5V, 3.3V, 5V Adjustable from 1.25V to 5.0V
- Power OK Output
- ♦ Low 125µA Ground Current
- 0.1µA Shutdown Current
- Thermal Overload Protection
- Output Current Limit
- Tiny 800mW 6-Pin SOT23 Package

Guaranteed 500mA Output Current

Low 120mV Dropout at 500mA

Ordering Information

PART*	TEMP RANGE	PIN-PACKAGE
MAX1818EUAT	-40°C to +85°C	6 SOT23-6
** *** * * *		

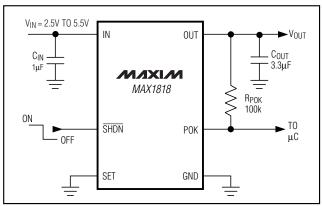
*Insert the desired two-digit suffix (see Selector Guide) into the blanks to complete the part number.

Note: The MAX1818 requires a special solder temperature profile described in the Absolute Maximum Ratings section.

_Selector Guide

PART AND SUFFIX	Vout	SOT23 TOP MARK
MAX1818EUT15	1.5V or Adj	AASO
MAX1818EUT18	1.8V or Adj	AANU
MAX1818EUT20	2.0V or Adj	AANV
MAX1818EUT25	2.5V or Adj	AANF
MAX1818EUT33	3.3V or Adj	AANG
MAX1818EUT50	5.0V or Adj	AANH

Typical Operating Circuit



_ Maxim Integrated Products 1

For pricing, delivery, and ordering information, please contact Maxim/Dallas Direct! at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

ABSOLUTE MAXIMUM RATINGS

IN, SHDN, POK, SET to GND0.3V	to +6V
OUT to GND0.3V to (VIN -	
Output Short-Circuit Duration	1min
Continuous Power Dissipation ($T_A = +70^{\circ}C$) (Note 1)	
6-Pin SOT23 (derate 10mW/°C above +70°C)8	800mW

Operating Temperature Range	40°C to +85°C
Junction Temperature	+150°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (soldering, 10s) (Note 2)+300°C

Note 1: Thermal properties are specified with product mounted on PC board with one square-inch of copper area and still air. With minimal copper, the SOT23 package dissipates 712mW at +70°C. With a quarter square inch of copper, it will dissipate 790mW at +70°C. Copper should be equally shared between the IN, OUT, and GND pins.

Note 2: This device is constructed using a unique set of packaging techniques that imposes a limit on the thermal profile to which the device can be exposed during board-level solder attach and rework. The limit permits only the use of the solder profiles recommended in the industry standard specification, IPC JEDEC-J-STD-020A, paragraph 7.6, Table 3 for the IR/VPR and convection reflow. Preheating is required. Hand or wave soldering is not allowed.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

 $(V_{IN} = V_{OUT} + 500mV \text{ or } V_{IN} = +2.5V$, whichever is greater; $\overline{SHDN} = IN$, $T_A = -40^{\circ}C$ to $+85^{\circ}C$, unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.) (Note 3)

PARAMETER	SYMBOL	CON	DITIONS	MIN	ТҮР	МАХ	UNITS
Input Voltage	V _{IN}			2.5		5.5	V
Input Undervoltage Lock-Out	Vuvlo	Rising, 75mV hystere	esis	2.0	2.15	2.3	V
		I _{OUT} = 100mA, T _A =	+25°C, V _{OUT} ≥ 2.0V	-1		+1	
		IOUT = 100mA, VOUT	r < 2.0V	-1.5		+1.5	
Output Voltage Accuracy (Preset Mode)	VOUT	$I_{OUT} = 100 \text{mA}, T_A =$	-40°C to +85°C	-2.5		+2.5	%
(Freset Mode)		$I_{OUT} = 1$ mA to 500m $T_A = -40$ °C to +85°C	NA, V _{IN} > V _{OUT} + 0.5V,	-3		+3	
Adjustable Output Voltage Range				1.25		5	V
SET Voltage Threshold		$V_{IN} = 2.7V,$	$T_A = 0^{\circ}C$ to +85°C	1.225	1.250	1.275	
(Adjustable Mode)	VSET	I _{OUT} = 100mA, V _{OUT} set to 2.0V	$T_A = -40^{\circ}C \text{ to } +85^{\circ}C$	1.213		1.288	V
Guaranteed Output Current (RMS)	IOUT	$V_{IN} \ge 2.7V$		500			mA
Short-Circuit Current Limit	ILIM	$V_{OUT} = 0, V_{IN} \ge 2.7 V_{OUT}$	1	0.55	0.8	1.8	А
In-Regulation Current Limit		V _{OUT} > 96% of nomi	nal value, V _{IN} ≥ 2.7V		1.6		А
SET Dual Mode™ Threshold				50	100	150	mV
SET Input Bias Current	ISET	V _{SET} = 1.25V		-100		+100	nA
Ground-Pin Current	1-	I _{OUT} = 100μA			125	250	
Ground-Pin Current	lq	I _{OUT} = 500mA			140		μA
			Vout = 5V		100	220	
Dropout Voltage (Note 4)	V _{IN} - V _{OUT}	$I_{OUT} = 500 \text{mA}$	V _{OUT} = 3.3V		120	215	mV
			V _{OUT} = 2.5V		210	360	
Line Regulation	ΔV_{LNR}	V _{IN} from (V _{OUT} + 10 I _{LOAD} = 5mA	0mV) to 5.5V,	-0.15	0	+0.15	%/V

Dual Mode is a trademark of Maxim Integrated Products



ELECTRICAL CHARACTERISTICS (continued)

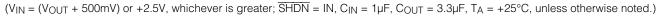
(VIN = V_{OUT} + 500mV or V_{IN} = +2.5V, whichever is greater; SHDN = IN, T_A = -40°C to +85°C, unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.) (Note 3)

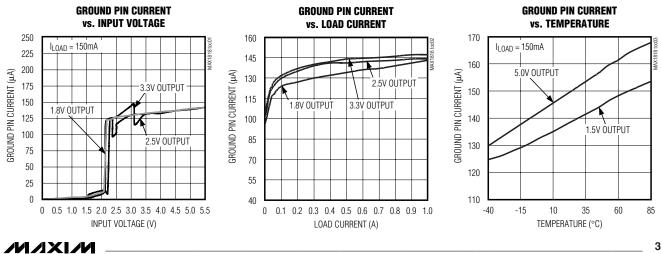
PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	МАХ	UNITS
Load Regulation	ΔV_{LDR}	I _{OUT} = 1mA to 500mA		0.4	1.0	%
Output Voltage Noise		10Hz to 1MHz, $C_{OUT} = 3.3 \mu F (ESR < 0.1 \Omega)$		115		μV _{RMS}
SHUTDOWN						
Shutdown Supply Current	IOFF	$\overline{\text{SHDN}} = \text{GND}, \text{V}_{\text{IN}} = 5.5\text{V}$		0.1	15	μΑ
	VIH	2.5V < V _{IN} < 5.5V	1.6			v
SHDN Input Threshold	VIL	2.5V < V _{IN} < 5.5V			0.6	v
		$\overline{SHDN} = IN \text{ or } GND, T_A = +25^{\circ}C$		1	25	~^
SHDN Input Bias Current	SHDN	$T_A = +85^{\circ}C$		5		nA
ΡΟΚ ΟυΤΡυΤ						
POK Output Low Voltage	VOL	POK sinking 1mA		0.01	0.1	V
Operating Voltage Range for Valid POK		POK sinking 100µA	1.0		5.5	V
POK Output High Leakage		POK = 5.5V, T _A = +25°C		1	25	,
Current		T _A = +85°C		5		nA
POK Threshold		Rising edge, referred to VOUT(NOMINAL)	90	93	96	%
THERMAL PROTECTION	•	· · · · · · · · · · · · · · · · · · ·	-			•
Thermal Shutdown Temperature	TSHDN			170		°C
Thermal Shutdown Hysteresis	ΔT_{SHDN}			20		°C

Note 3: All devices are 100% production tested at T_A = +25°C. Limits over the operating temperature range are guaranteed by design.

Note 4: The Dropout Voltage is defined as VIN - VOUT, when VOUT is 100mV below the value of VOUT measured for VIN = VOUT(NOM) + 500mV. Since the minimum input voltage is 2.5V, this specification is only meaningful when VOUT(NOM) > 2.5V. For V_{OUT(NOM)} between 2.5V and 3.5V, use the following equations: Typical Dropout = -93mV/V × V_{OUT(NOM)} + 445mV; Guaranteed Maximum Dropout = -137mV/V × V_{OUT(NOM)} +7004mV. For V_{OUT(NOM)} > 3.5V, Typical Dropout = 120mV; Maximum Dropout = 220mV.

Typical Operating Characteristics





3.0

25

2.0

1.5

1.0

0.5

0

Typical Operating Characteristics (continued)

3.0

2.5

2.0

1.5

1.0

0.5

0

-0.5

-1.0

-1.5

-2.0

-2.5

-3.0

OUTPUT VOLTAGE ACCURACY (%)

 $(V_{IN} = (V_{OUT} + 500 \text{mV}) \text{ or } +2.5 \text{V}$, whichever is greater; $\overline{\text{SHDN}} = \text{IN}$, $C_{IN} = 1 \mu \text{F}$, $C_{OUT} = 3.3 \mu \text{F}$, $T_A = +25 \text{°C}$, unless otherwise noted.)

OUTPUT VOLTAGE ACCURACY

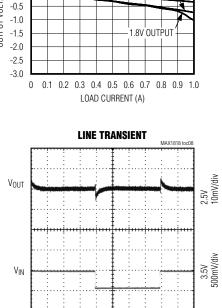
vs. LOAD CURRENT

2.5V OUTPUT

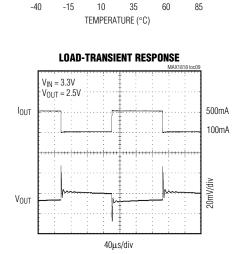
3.3V OUTPUT

vs. LOAD CURRENT 180 $V_{OUT} = 3.3V$ Т_А = +85°С 150 (%) OUTPUT VOLTAGE ACCURACY DROPOUT VOLTAGE (mV) 120 90 $T_{\Delta} = +25^{\circ}C$ 60 T_A = -40°C 30 0 0 0.1 0.2 0.3 0.4 0.5 LOAD CURRENT (A) **POWER-SUPPLY REJECTION RATIO** vs. FREQUENCY 80 70 $C_{OUT} = 3.3 \mu F$ $V_{OUT} = 3.3 V$ 60 50 $R_{LOAD} = 100\Omega$

DROPOUT VOLTAGE



100µs/div



OUTPUT VOLTAGE ACCURACY

vs. TEMPERATURE

ILOAD = 150mA

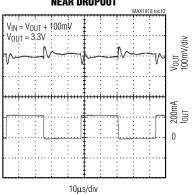
V_{OUT} = 1.5V TO 5.0V

LOAD-TRANSIENT RESPONSE NEAR DROPOUT

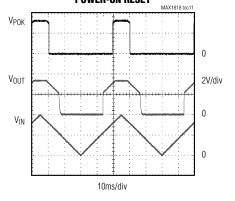
100

1000

10



POWER-ON RESET



PSRR (dB)

40 30

20

10 0

0.01

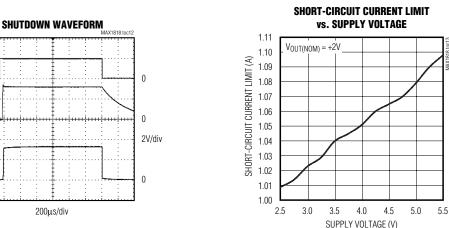
0.1

1

FREQUENCY (kHz)

Typical Operating Characteristics (continued)

 $(V_{IN} = (V_{OUT} + 500mV) \text{ or } +2.5V, \text{ whichever is greater; } \overline{SHDN} = IN, C_{IN} = 1\mu F, C_{OUT} = 3.3\mu F, T_A = +25^{\circ}C, \text{ unless otherwise noted.})$



MAX1818

Pin Description

PIN	NAME	FUNCTION
1	IN	Regulator Input. Supply voltage can range from 2.5V to 5.5V. Bypass with a 1μ F capacitor to GND (see <i>Capacitor Selection and Regulator Stability</i>).
2	РОК	Open-Drain POK Output. POK remains low while the output voltage (V_{OUT}) is below the POK threshold. Connect a 100k Ω pull-up resistor from POK to OUT to obtain an output voltage.
3	SHDN	Active-Low Shutdown Input. A logic low reduces supply current below 15µA. In shutdown, the POK output is low, and OUT is high impedance. Connect to IN for normal operation.
4	GND	Ground
5	SET	Voltage-Setting Input. Connect to GND for preset output. Connect to a resistive voltage-divider between OUT and GND to set the output voltage between 1.25V and 5.0V.
6	OUT	Regulator Output. Sources up to 500mA. Bypass with a 3.3μ F low-ESR capacitor to GND. Use a 4.7μ F capacitor for output voltages below 2V.

Detailed Description

The MAX1818 is a low-dropout, low-quiescent-current linear regulator designed primarily for battery-powered applications. The device supplies loads up to 500mA and is available with preset output voltages. As illustrated in Figure 1, the MAX1818 consists of a 1.25V reference, error amplifier, P-channel pass transistor, and internal feedback voltage-divider.

The 1.25V reference is connected to the error amplifier, which compares this reference with the feedback voltage and amplifies the difference. If the feedback voltage is lower than the reference voltage, the

pass-transistor gate is pulled lower, which allows more current to pass to the output and increases the output voltage. If the feedback voltage is too high, the passtransistor gate is pulled up, allowing less current to pass to the output.

The output voltage is fed back through either an internal resistive divider connected to OUT or an external resistor network connected to SET. The Dual Mode comparator examines V_{SET} and selects the feedback path. If V_{SET} is below 50mV, the internal feedback path is used and the output is regulated to the factory-preset voltage.

<u>///XI///</u>

SHUTDOWN

VOLTAGE

VOUT

POK

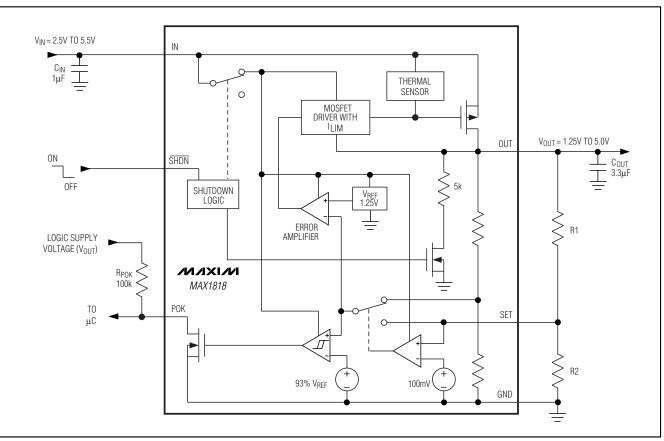


Figure 1. Functional Diagram

Additional blocks include an output current limiter, thermal sensor, and shutdown logic.

Internal P-Channel Pass Transistor

The MAX1818 features a 0.25 Ω P-channel MOSFET pass transistor. Unlike similar designs using PNP pass transistors, P-channel MOSFETs require no base drive, which reduces quiescent current. PNP-based regulators also waste considerable current in dropout when the pass transistor saturates, and use high base-drive currents under large loads. The MAX1818 does not suffer from these problems and consumes only 125 μ A of quiescent current under heavy loads as well as in dropout.

Output Voltage Selection

The MAX1818's Dual Mode operation allows operation in either a preset voltage mode or an adjustable mode. Connect SET to GND to select the preset output voltage. The two-digit part number suffix identifies the output voltage (see *Selector Guide*). For example, the MAX1818EUT33 has a preset 3.3V output voltage.

The output voltage may also be adjusted by connecting a voltage-divider from OUT to SET to GND (Figure 2). Select R2 in the 25k Ω to 100k Ω range. Calculate R1 with the following equation:

$$R1 = R2 [(V_{OUT} / V_{SET}) - 1]$$

where V_{SET} = 1.25V, and V_{OUT} may range from 1.25V to 5.0V.

Shutdown

Pull \overline{SHDN} low to enter shutdown. During shutdown, the output is disconnected from the input and supply current drops to 0.1µA. When in shutdown, POK pulls low and OUT is high impedance. The capacitance and load at OUT determine the rate at which V_{OUT} decays. SHDN can be pulled as high as 6V, regardless of the input and output voltage.

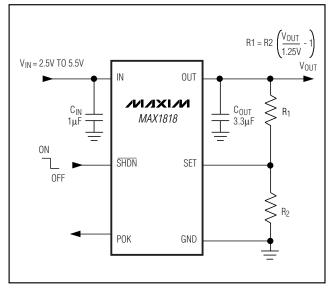


Figure 2. Adjustable Output Using External Feedback Resistors

POK Output The power OK (POK) output pulls low when OUT is less than 93% of the nominal regulation voltage. Once OUT exceeds 93% of the nominal voltage, POK goes high impedance. POK is an open-drain N-channel output. To obtain a voltage output, connect a pullup resistor from POK to OUT. A 100k Ω resistor works well for most applications. POK can be used as a power-okay (POK) signal to a microcontroller (μ C), or drive an external LED to indicate power failure. When the MAX1818 is shut down, POK is held low independent of the output voltage. If unused, leave POK grounded or unconnected.

Current Limit

The MAX1818 monitors and controls the pass transistor's gate voltage, limiting the output current to 0.8A (typ). This current limit doubles when the output voltage is within 4% of the nominal value to improve performance with large load transients.

Thermal Overload Protection

Thermal overload protection limits total power dissipation in the MAX1818. When the junction temperature exceeds $T_J = +170^{\circ}$ C, a thermal sensor turns off the pass transistor, allowing the IC to cool. The thermal sensor turns the pass transistor on again after the junction temperature cools by 20°C, resulting in a pulsed output during continuous thermal overload conditions. Thermal overload protection protects the MAX1818 in



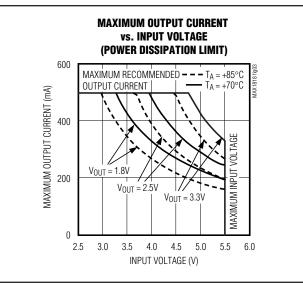


Figure 3. Power Operating Regions: Maximum Output vs. Supply Voltage

the event of fault conditions. For continuous operation, do not exceed the absolute maximum junction-temperature rating of T_J = +150°C.

Operating Region and Power Dissipation The MAX1818's maximum power dissipation depends on the thermal resistance of the IC package and circuit board, the temperature difference between the die junction and ambient air, and the rate of air flow. The power dissipated in the device is $P = I_{OUT} \times (V_{IN} - V_{OUT})$. The maximum allowed power dissipation is 800mW at $T_A = +70^{\circ}$ C or:

$$P_{MAX} = (T_{J(MAX)} - T_A) / (\theta_{JC} + \theta_{CA})$$

where T_J - T_A is the temperature difference between the MAX1818 die junction and the surrounding air, θ_{JC} is the thermal resistance of the junction to the case, and θ_{CA} is the thermal resistance from the case through the PC board, copper traces, and other materials to the surrounding air. For best heatsinking, the copper area should be equally shared between the IN, OUT, and GND pins.

The MAX1818 delivers up to 0.5A RMS and operates with input voltages up to +5.5V, but not simultaneously. High output currents can only be sustained when input-output differential voltages are low, as shown in Figure 3.

MAX1818

Applications Information

Capacitor Selection and Regulator Stability

Capacitors are required at the MAX1818's input and output for stable operation over the full temperature range and with load currents up to 500mA. Connect a 1 μ F capacitor between IN and ground and a 3.3 μ F low-ESR capacitor between OUT and ground. For output voltages less than 2V, use a 4.7 μ F low-ESR output capacitor. The input capacitor (CIN) lowers the source impedance of the input supply. Reduce noise and improve load-transient response, stability, and power-supply rejection by using larger output capacitors, such as 10 μ F.

The output capacitor's (C_{OUT}) equivalent series resistance (ESR) affects stability and output noise. Use output capacitors with an ESR of 0.1 Ω or less to ensure stability and optimum transient response. Surfacemount ceramic capacitors have very low ESR and are commonly available in values up to 10µF. Connect C_{IN} and C_{OUT} as close to the MAX1818 as possible to minimize the impact of PC board trace inductance.

Noise, PSRR, and Transient Response

The MAX1818 is designed to operate with low dropout voltages and low quiescent currents in battery-powered systems while still maintaining good noise, transient response, and AC rejection. See the Typical Operating Characteristics for a plot of power-supply rejection ratio (PSRR) versus frequency. When operating from noisy sources, improved supply-noise rejection and transient response can be achieved by increasing the values of

the input and output bypass capacitors and through passive filtering techniques.

The MAX1818 load-transient response (see Typical Operating Characteristics) shows two components of the output response: a DC shift from the output impedance due to the load current change, and the transient response. A typical transient response for a step change in the load current from 100mA to 500mA is 8mV. Increasing the output capacitor's value and decreasing the ESR attenuates the overshoot.

Input-Output (Dropout) Voltage

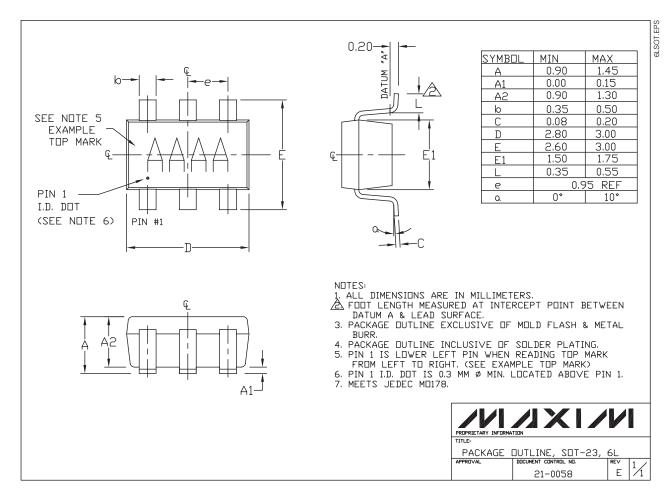
A regulator's minimum input-to-output voltage differential (dropout voltage) determines the lowest usable supply voltage. In battery-powered systems, this determines the useful end-of-life battery voltage. Because the MAX1818 uses a P-channel MOSFET pass transistor, its dropout voltage is a function of drain-tosource on-resistance (R_{DS(ON)}) multiplied by the load current (see *Typical Operating Characteristics*).

The MAX1818 ground current remains below 150 μA in dropout.

Chip Information

TRANSISTOR COUNT: 845

Package Information



Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time.

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/N/XI/N				SITE	PART NC SEARCH
HAT'S NEW PRODUCTS	SOLUTIONS	S DES	SIGN APPNOTES SUPPORT	BUY	COMPANY MEMBERS
			MAX1818 Part Number Table		
Notes:					
 full data sheet (PI 2. Other options and 3. Didn't Find What 's business day. 4. Part number suffi sheet or Part Nam 	DF, 268kB). I links for pur You Need? As xes: T or T&F ning Convent	chasing p k our app R = tape a ions.	et for further information on this produ parts are listed at: http://www.maxim plications engineers. Expert assistance and reel; + = RoHS/lead-free; # = Ro d on the drawing. "PkgCode/Variation	-ic.com/sales in finding parts oHS/lead-exemp	s, usually within one ot. More: See full data
Part Number	Free Sample	Buy Direct	Package: TYPE PINS SIZE DRAWING CODE/VAR *	Temp	RoHS/Lead-Free? Materials Analysis
MAX1818EUT25#G16			SOT-23;6 pin; Dwg: 21-0058I (PDF) Use pkgcode/variation: U6FH-6*	-40C to +85C	RoHS/Lead-Free: Yes Materials Analysis
MAX1818EUT15#TG16			SOT-23;6 pin; Dwg: 21-0058I (PDF) Use pkgcode/variation: U6FH-6*	-40C to +85C	RoHS/Lead-Free: Yes Materials Analysis
MAX1818EUT18#TG16			SOT-23;6 pin; Dwg: 21-0058I (PDF) Use pkgcode/variation: U6FH-6*	-40C to +85C	RoHS/Lead-Free: Yes Materials Analysis
MAX1818EUT20#TG16			SOT-23;6 pin; Dwg: 21-0058I (PDF) Use pkgcode/variation: U6FH-6*	-40C to +85C	RoHS/Lead-Free: Yes Materials Analysis
MAX1818EUT25#TG16			SOT-23;6 pin; Dwg: 21-0058I (PDF) Use pkgcode/variation: U6FH-6*	-40C to +85C	RoHS/Lead-Free: Yes Materials Analysis
MAX1818EUT33#TG16			SOT-23;6 pin; Dwg: <mark>21-0058I</mark> (PDF) Use pkgcode/variation: U6FH-6*	-40C to +85C	RoHS/Lead-Free: Yes Materials Analysis
MAX1818EUT50#TG16			SOT-23;6 pin; Dwg: 21-0058I (PDF) Use pkgcode/variation: U6FH-6*	-40C to +85C	RoHS/Lead-Free: Yes Materials Analysis

MAX1818EUT15#G16	SOT-23;6 pin; Dwg: 21-0058I (PDF) Use pkgcode/variation: U6FH-6*	-40C to +85C	RoHS/Lead-Free: Yes Materials Analysis
MAX1818EUT50#G16	SOT-23;6 pin; Dwg: 21-0058I (PDF) Use pkgcode/variation: U6FH-6*	-40C to +85C	RoHS/Lead-Free: Yes Materials Analysis
MAX1818EUT33#G16	SOT-23;6 pin; Dwg: 21-0058I (PDF) Use pkgcode/variation: U6FH-6*	-40C to +85C	RoHS/Lead-Free: Yes Materials Analysis
MAX1818EUT20#G16	SOT-23;6 pin; Dwg: 21-0058I (PDF) Use pkgcode/variation: U6FH-6*	-40C to +85C	RoHS/Lead-Free: Yes Materials Analysis
MAX1818EUT50	SOT-23;6 pin; Dwg: 21-0058I (PDF) Use pkgcode/variation: U6F-6*	-40C to +85C	RoHS/Lead-Free: No Materials Analysis
MAX1818EUT33	SOT-23;6 pin; Dwg: 21-0058I (PDF) Use pkgcode/variation: U6F-6*	-40C to +85C	RoHS/Lead-Free: No Materials Analysis
MAX1818EUT25	SOT-23;6 pin; Dwg: 21-0058I (PDF) Use pkgcode/variation: U6F-6*	-40C to +85C	RoHS/Lead-Free: No Materials Analysis
MAX1818EUT20	SOT-23;6 pin; Dwg: 21-0058I (PDF) Use pkgcode/variation: U6F-6*	-40C to +85C	RoHS/Lead-Free: No Materials Analysis
MAX1818EUT18	SOT-23;6 pin; Dwg: 21-0058I (PDF) Use pkgcode/variation: U6F-6*	-40C to +85C	RoHS/Lead-Free: No Materials Analysis
MAX1818EUT15	SOT-23;6 pin; Dwg: 21-0058I (PDF) Use pkgcode/variation: U6F-6*	-40C to +85C	RoHS/Lead-Free: No Materials Analysis
MAX1818EUT18-T	SOT-23;6 pin; Dwg: 21-0058I (PDF) Use pkgcode/variation: U6F-6*	-40C to +85C	RoHS/Lead-Free: No Materials Analysis
MAX1818EUT20-T	SOT-23;6 pin; Dwg: 21-0058I (PDF) Use pkgcode/variation: U6F-6*	-40C to +85C	RoHS/Lead-Free: No Materials Analysis
MAX1818EUT25-T	SOT-23;6 pin; Dwg: 21-0058I (PDF) Use pkgcode/variation: U6F-6*	-40C to +85C	RoHS/Lead-Free: No Materials Analysis
MAX1818EUT33-T	SOT-23;6 pin; Dwg: 21-0058I (PDF) Use pkgcode/variation: U6F-6*	-40C to +85C	RoHS/Lead-Free: No Materials Analysis
MAX1818EUT50-T	SOT-23;6 pin; Dwg: 21-0058I (PDF) Use pkgcode/variation: U6F-6*	-40C to +85C	RoHS/Lead-Free: No Materials Analysis
MAX1818EUT18-TG16	SOT-23;6 pin; Dwg: 21-0058I (PDF) Use pkgcode/variation: U6F-6*	-40C to +85C	RoHS/Lead-Free: No Materials Analysis

MAX1818EUT15-T	SOT-23;6 pin; Dwg: 21-0058I (PDF) Use pkgcode/variation: U6F-6*	-40C to +85C	RoHS/Lead-Free: No Materials Analysis
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